

Alma Mater Studiorum Università di Bologna
Archivio istituzionale della ricerca

A High-Gain Nonlinear Observer with Limited Gain Power

This is the final peer-reviewed author's accepted manuscript (postprint) of the following publication:

Published Version:

A High-Gain Nonlinear Observer with Limited Gain Power / Astolfi, Daniele; Marconi, Lorenzo. - In: IEEE TRANSACTIONS ON AUTOMATIC CONTROL. - ISSN 0018-9286. - STAMPA. - 60:11(2015), pp. 7054521.3059-7054521.3064. [10.1109/TAC.2015.2408554]

Availability:

This version is available at: <https://hdl.handle.net/11585/555935> since: 2016-07-18

Published:

DOI: <http://doi.org/10.1109/TAC.2015.2408554>

Terms of use:

Some rights reserved. The terms and conditions for the reuse of this version of the manuscript are specified in the publishing policy. For all terms of use and more information see the publisher's website.

This item was downloaded from IRIS Università di Bologna (<https://cris.unibo.it/>).
When citing, please refer to the published version.

(Article begins on next page)

This is the final peer-reviewed accepted manuscript of:

*D. Astolfi and L. Marconi, "A **High-Gain Nonlinear Observer With Limited Gain Power**" in IEEE Transactions on Automatic Control, vol. 60, no. 11, pp. 3059-3064, Nov. 2015*

The final published version is available online at:

<https://doi.org/10.1109/TAC.2015.2408554>

Rights / License:

The terms and conditions for the reuse of this version of the manuscript are specified in the publishing policy. For all terms of use and more information see the publisher's website.

This item was downloaded from IRIS Università di Bologna (<https://cris.unibo.it/>)

When citing, please refer to the published version.

A High-Gain Nonlinear Observer with Limited Gain Power

Daniele Astolfi and Lorenzo Marconi

Abstract—In this note we deal with a new observer for nonlinear systems of dimension n in canonical observability form. We follow the standard high-gain paradigm, but instead of having an observer of dimension n with a gain that grows up to power n , we design an observer of dimension $2n - 2$ with a gain that grows up only to power 2.

Index Terms—Observability, nonlinear observers, high-gain observers.

I. INTRODUCTION

In this note we consider the problem of state observation for nonlinear systems of the form

$$\dot{z} = f(z) + \bar{d}(t), \quad y = h(z) + \nu(t) \quad (1)$$

where $z \in \mathcal{Z} \subseteq \mathbb{R}^n$ is the state, $y \in \mathbb{R}$ is the measured output, $f(\cdot)$ and $h(\cdot)$ are sufficiently smooth functions, $\bar{d}(t) \in \mathbb{R}^n$ is a bounded disturbance and $\nu \in \mathbb{R}$ is the measurement noise. Among the different techniques for observer design available in literature (see [2], [4]) we are particularly interested to the so-called high-gain methods that have been shown to be effective in many control scenarios. In this respect we assume that the pair $(f(\cdot), h(\cdot))$ fulfils a *uniform observability assumption* (see Definition 1.2 in [2]), which implies the existence of a diffeomorphism $\phi : \mathcal{Z} \rightarrow \mathbb{R}^n$ such that the dynamics of the new state variable $x = \phi(z)$ is described by the *canonical observability form* (see Theorem 4.1 in [2])

$$\dot{x} = A_n x + B_n \varphi(x) + d(x, t), \quad y = C_n x + \nu(t) \quad (2)$$

where $\varphi(\cdot)$ is a locally Lipschitz function,

$$d(x, t) := \left. \frac{d\phi(z)}{dz} \right|_{z=\phi_{\text{inv}}(x)} \bar{d}(t),$$

with $\phi_{\text{inv}}(\cdot)$ the inverse of $\phi(\cdot)$ (namely $\phi_{\text{inv}} \circ \phi(z) = z$ for all $z \in \mathcal{Z}$), and (A_n, B_n, C_n) is a triplet in “prime form” of dimension n , that is

$$\begin{aligned} A_n &= \begin{pmatrix} 0_{(n-1) \times 1} & I_{n-1} \\ 0 & 0_{1 \times (n-1)} \end{pmatrix} & B_n &= \begin{pmatrix} 0_{(n-1) \times 1} \\ 1 \end{pmatrix} \\ C_n &= \begin{pmatrix} 1 & 0_{1 \times (n-1)} \end{pmatrix}. \end{aligned} \quad (3)$$

System (2) is defined on a set $\mathcal{X} := \phi(\mathcal{Z}) \subseteq \mathbb{R}^n$.

For the class of systems (2), it is a well-known fact ([3]) that the problem of asymptotically, in case $d(x, t) \equiv 0$ and

$\nu(t) \equiv 0$, estimating the state x can be addressed by means of a *high-gain* nonlinear observer of the form

$$\dot{\hat{x}} = A_n \hat{x} + B_n \varphi_s(\hat{x}) + D_n(\ell) K_n (C_n \hat{x} - y) \quad (4)$$

with

$$D_n(\ell) = \text{diag}(\ell, \dots, \ell^n), \quad K_n = \begin{pmatrix} c_1 & \dots & c_n \end{pmatrix}^\top,$$

where ℓ is a high-gain design parameter taken sufficiently large (i.e. $\ell \geq \ell^*$ with $\ell^* \geq 1$), the c_i ’s are chosen so that the matrix $(A + KC)$ is Hurwitz (i.e. all its eigenvalues are on the left-half complex plane), and $\varphi_s(\cdot) : \mathbb{R}^n \rightarrow \mathbb{R}$ is an appropriate “saturated” version of $\varphi(\cdot)$. As a matter of fact, it can be proved that if $d(x, t) \equiv 0$ and $\nu(t) \equiv 0$, if $\varphi(\cdot)$ is uniformly Lipschitz in \mathcal{X} , namely there exists a $\bar{\varphi} > 0$ such that

$$\|\varphi(x') - \varphi(x'')\| \leq \bar{\varphi} \|x' - x''\| \quad \forall x', x'' \in \mathcal{X}, \quad (5)$$

and $\varphi_s(\cdot)$ is chosen bounded and to agree with $\varphi(\cdot)$ on \mathcal{X} , the observation error $e(t) = x(t) - \hat{x}(t)$ originating from (2) and (4) exponentially converges to the origin with an exponential decay rate of the form

$$\|x(t) - \hat{x}(t)\| \leq \alpha \ell^{n-1} \exp\left(-\frac{\beta}{\ell} t\right)$$

where α and β are positive constants, for all possible initial condition $\hat{x}(0) \in \mathbb{R}^n$ as long as $x(t) \in \mathcal{X}$. In particular, note that the exponential decay rate may be arbitrarily assigned by the value of ℓ with a polynomial “peaking” in ℓ of order $n-1$. It is worth noting that the uniform Lipschitz condition (5) is automatically fulfilled if \mathcal{X} is a compact set. In case $d(x, t)$ or $\nu(t)$ are not identically zero, as long as they are bounded for all $t \geq 0$ and for all¹ $x \in \mathcal{X}$, the observer (4) guarantees a bound on the estimation error that depends on the bound of $d(\cdot, \cdot)$, of $\nu(\cdot)$ and on the value of ℓ . In particular, the following asymptotic bounds can be proved

$$\limsup_{t \rightarrow \infty} \|x(t) - \hat{x}(t)\| \leq \gamma \max \left\{ \limsup_{t \rightarrow \infty} \left\| \frac{1}{\ell} \Gamma(\ell) d(x(t), t) \right\|, \limsup_{t \rightarrow \infty} \|\ell^{n-1} \nu(t)\| \right\}$$

where γ is a positive constant and

$$\Gamma(\ell) = \text{diag}(\ell^{n-1}, \dots, \ell, 1). \quad (6)$$

As above the previous asymptotic bound holds for all possible $\hat{x}(0) \in \mathbb{R}^n$ as long as $x(t) \in \mathcal{X}$. Note that a high value of ℓ leads to an arbitrarily small asymptotic gain on the n -th

D. Astolfi is with CASY-DEI, University of Bologna, Italy and with the École des Mines de Paris, Fontainebleau, France, daniele.astolfi@unibo.it. L. Marconi is with the CASY-DEI, University of Bologna, Italy, lorenzo.marconi@unibo.it. Work supported by the European project SHERPA (ICT 600958).

¹Boundedness of $d(x, t)$ is automatically guaranteed if \mathcal{X} is compact. This, in turn, is the typical case when such observers are used in semiglobal output feedback stabilisation problems, [5].

component of the disturbance $d(\cdot, \cdot)$. On the other hand, a large value of ℓ is, in general, detrimental for the sensitivity of the asymptotic estimate to the sensor noise and to the first $n - 2$ disturbance components.

Observers of the form (4) are routinely used in many observation and control problems. For instance, the feature of having an exponential decay rate and an asymptotic bound on the last component of d that can be arbitrarily imposed by the value of ℓ is the main reason why the above observer plays a fundamental role in output feedback stabilisation and in setting up semiglobal nonlinear separation principles ([5], [6]). In that case the set \mathcal{X} is an arbitrarily large compact set which is made invariant by the design of the state feedback stabilisation law and of the high-gain observer. We observe that, although the asymptotic gain with respect to ν increase with ℓ^{n-1} , the observer is anyway able to guarantee ISS with respect to the sensor noise ([7]).

The main drawback of observers of the form (4), though, is related to the increasing power (up to the order n) of the high-gain parameter ℓ , which makes the practical numerical implementation an hard task when n or ℓ are very large. Motivated by these considerations, in this note we propose a new observer for the class of systems (2) that preserves the same high-gain features of (4) but which substantially overtakes the implementation problems due to the high-gain powered up to the order n . Specifically, we present a high-gain observer structure with a gain which grows only up to power 2 (instead of n), at the price of having the observer state dimension $2n - 2$ instead of n .

II. MAIN RESULT

We start by presenting a technical lemma instrumental to the proof of the main result presented in Proposition 1. Let $E_i \in \mathbb{R}^{2 \times 2}$, $Q_i \in \mathbb{R}^{2 \times 2}$, $i = 1, \dots, n - 1$, and $N \in \mathbb{R}^{2 \times 2}$ be matrices defined as

$$E_i = \begin{pmatrix} -k_{i1} & 1 \\ -k_{i2} & 0 \end{pmatrix}, \quad Q_i = \begin{pmatrix} 0 & k_{i1} \\ 0 & k_{i2} \end{pmatrix}, \quad N = \begin{pmatrix} 0 & 0 \\ 0 & 1 \end{pmatrix}$$

where (k_{i1}, k_{i2}) are positive coefficients, and let $M \in \mathbb{R}^{(2n-2) \times (2n-2)}$ be the block-tridiagonal matrix defined as

$$M = \begin{pmatrix} E_1 & N & 0 & \dots & \dots & 0 \\ Q_2 & E_2 & N & \ddots & & \vdots \\ 0 & \ddots & \ddots & \ddots & \ddots & \vdots \\ \vdots & \ddots & Q_i & E_i & N & \ddots \\ \vdots & & \ddots & \ddots & \ddots & \ddots \\ \vdots & & & \ddots & Q_{n-2} & E_{n-2} & N \\ 0 & \dots & \dots & \dots & 0 & Q_{n-1} & E_{n-1} \end{pmatrix}. \quad (7)$$

It turns out that the eigenvalues of M can be arbitrarily assigned by appropriately choosing the coefficients (k_{i1}, k_{i2}) , $i = 1, \dots, n - 1$, as claimed in the next lemma.

Lemma 1 *Let $\mathcal{P}(\lambda) = \lambda^{2n-2} + m_1 \lambda^{2n-3} + \dots + m_{2n-3} \lambda + m_{2n-2}$ be an arbitrary Hurwitz polynomial. There exists a*

choice of (k_{i1}, k_{i2}) , $i = 1, \dots, n - 1$, such that the characteristic polynomial of M coincides with $\mathcal{P}(\lambda)$.

The proof of this Lemma is deferred to the appendix where a constructive procedure for designing (k_{i1}, k_{i2}) given $\mathcal{P}(\lambda)$ is presented.

The structure of the proposed observer has the following form

$$\begin{aligned} \dot{\xi}_i &= A \xi_i + N \xi_{i+1} + D_2(\ell) K_i e_i & i = 1, \dots, n-2 \\ &\vdots \\ \dot{\xi}_{n-1} &= A \xi_{n-1} + B \varphi_s(\hat{x}') + D_2(\ell) K_{(n-1)} e_{n-1} \end{aligned} \quad (8)$$

where (A, B, C) is a triplet in prime form of dimension 2, $\xi_i \in \mathbb{R}^2$, $K_i = (k_{i1} \ k_{i2})^T$, $D_2(\ell) = \text{diag}(\ell, \ell^2)$,

$$\hat{x}' = L_1 \xi \quad L_1 = \text{blkdiag}(\underbrace{C, \dots, C}_{(n-2) \text{ times}}, I_2)$$

$$\xi = \text{col}(\xi_1, \dots, \xi_{n-1}) \in \mathbb{R}^{2n-2},$$

$$e_1 = y - C \xi_1, \quad e_i = B^T \xi_{i-1} - C \xi_i \quad i = 2, \dots, n-1,$$

and $\varphi_s(\cdot)$ is an appropriate saturated version of $\varphi(\cdot)$.

The variable \hat{x}' represents an asymptotic estimate of the state x of (2). It is obtained by “extracting” n components from the state ξ according to the matrix L_1 defined above. As clarified next, the redundancy of the observer can be used to extract from ξ an extra state estimation that is

$$\hat{x}'' = L_2 \xi \quad L_2 = \text{blkdiag}(I_2, \underbrace{B^T, \dots, B^T}_{(n-2) \text{ times}}).$$

The following proposition shows that the observer (8) recovers the same asymptotic properties for the two estimates \hat{x}' and \hat{x}'' of the “standard” high-gain observer (4). In the statement of the proposition we let

$$\hat{\mathbf{x}} = \text{col}(\hat{x}', \hat{x}''), \quad \mathbf{x} = \text{col}(x, x).$$

Proposition 1 *Consider system (2) and the observer (8) with the coefficients (k_{i1}, k_{i2}) fixed so that the matrix M defined in (7) is Hurwitz (see Lemma 1). Let $\varphi_s(\cdot)$ be any bounded function that agrees with $\varphi(\cdot)$ on \mathcal{X} , and assume that $d(x, t)$ is bounded for all $x \in \mathcal{X}$ and for all $t \geq 0$. Then there exist $c_i > 0$, $i = 1, \dots, 4$, and $\ell^* \geq 1$ such that for any $\ell \geq \ell^*$ and for any $\xi(0) \in \mathbb{R}^{2n-2}$, the following bound holds*

$$\begin{aligned} \|\hat{\mathbf{x}}(t) - \mathbf{x}(t)\| &\leq \max\{c_1 \ell^{n-1} \exp(-\frac{c_2}{\ell} t) \|\hat{\mathbf{x}}(0) - \mathbf{x}(0)\|, \\ &c_3 \|\frac{1}{\ell} \Gamma(\ell) d(\cdot)\|_\infty, c_4 \ell^{n-1} \|\nu(\cdot)\|_\infty\} \end{aligned} \quad (9)$$

where $\Gamma(\ell)$ is as in (6), for all $t \geq 0$ such that $x(t) \in \mathcal{X}$.

Proof Consider the change of variable

$$\xi_i \mapsto \tilde{\xi}_i := \xi_i - \text{col}(x_i, x_{i+1})$$

by which system (8) transforms as

$$\begin{aligned}\dot{\tilde{\xi}}_1 &= H_1 \tilde{\xi}_1 + N \tilde{\xi}_2 - \bar{d}_1(t) + D_2(\ell) K_1 \nu(t) \\ \dot{\tilde{\xi}}_i &= H_i \tilde{\xi}_i + N \tilde{\xi}_{i+1} + D_2(\ell) K_i B^T \tilde{\xi}_{i-1} - \bar{d}_i(t) \\ &\quad i = 2, \dots, n-2 \\ \dot{\tilde{\xi}}_{n-1} &= H_{n-1} \tilde{\xi}_{n-1} + D_2(\ell) K_{n-1} B^T \tilde{\xi}_{n-2} + \\ &\quad B \Delta \varphi(\tilde{\xi}, x) - \bar{d}_{n-1}(t)\end{aligned}$$

where

$$H_i = A - D_2(\ell) K_i C, \quad \bar{d}_i(t) = (d_i(x(t), t), d_{i+1}(x(t), t))^T,$$

for $i = 1, \dots, n-1$, with $d_i(t)$ the i -th element of the vector $d(t)$, $\tilde{\xi} = \text{col}(\tilde{\xi}_1, \dots, \tilde{\xi}_{n-1})$, and

$$\Delta \varphi(\tilde{\xi}, x) = \varphi_s(L_1 \tilde{\xi} + x) - \varphi(x).$$

Rescale now the variables $\tilde{\xi}_i$ as follows

$$\tilde{\xi}_i \mapsto \varepsilon_i = \ell^{2-i} D_2(\ell)^{-1} \tilde{\xi}_i \quad i = 1, \dots, n-1.$$

By letting $\varepsilon = \text{col}(\varepsilon_1 \dots \varepsilon_{n-1})$, an easy calculation shows that

$$\dot{\varepsilon} = \ell M \varepsilon + \ell^{-(n-1)} (B_{2n-2} \Delta \varphi_\ell(\varepsilon, x) + v_\ell(t) + n_\ell(t)), \quad (10)$$

where B_{2n-2} is the zero column vector of dimension $2n-2$ with a 1 in the last position, and

$$\begin{aligned}v_\ell(t) &= -\text{col}(\ell^n D_2(\ell)^{-1} \bar{d}_1, \dots, \ell^{n-i+1} D_2(\ell)^{-1} \bar{d}_i, \dots, \\ &\quad \ell^2 D_2(\ell)^{-1} \bar{d}_{n-1}), \\ n_\ell(t) &= \ell^n \bar{K}_1 \nu(t), \quad \bar{K}_1 = \text{col}(K_1, 0, \dots, 0),\end{aligned}$$

and

$$\Delta \varphi_\ell(\varepsilon, x) = \varphi_s(S(\ell) \varepsilon + x) - \varphi(x),$$

where $S(\ell) = \text{diag}(\frac{1}{\ell}, 1, \ell, \dots, \ell^{n-3}) \otimes D_2(\ell)$. Being $\varphi(\cdot)$ uniformly Lipschitz in \mathcal{X} and $\varphi_s(\cdot)$ bounded, there exists a $\delta_1 > 0$, $\delta_2 > 0$ and $\delta_3 > 0$ such that

$$\begin{aligned}\|\ell^{-(n-1)} \Delta \varphi_\ell(\varepsilon, x)\| &\leq \delta_1 \|\varepsilon\|, \\ \|\ell^{-(n-1)} v_\ell(t)\| &\leq \delta_2 \|\ell D_n(\ell)^{-1} d(x(t), t)\|, \\ \|\ell^{-(n-1)} n_\ell(t)\| &\leq \delta_3 \|\ell \nu(t)\|,\end{aligned}$$

for all $\varepsilon \in \mathbb{R}^{2n-2}$, $x \in \mathcal{X}$ and $\ell \geq 1$. The rest of the proof follows standard Lyapunov arguments that, for sake of completeness, are briefly recalled. Let $P = P^T$ be such that $PM + M^T P = -I$ and consider the Lyapunov function $V = \varepsilon^T P \varepsilon$. Taking derivative of V along the solutions of (10), using the previous bounds and letting $\ell^* = 4\delta_1 \|P\|$, one obtains that there exist positive constants a_1, a_2, a_3 such that for any $\ell \geq \ell^*$

$$\begin{aligned}\|\varepsilon\| &\geq \max\{a_1 \|D_n(\ell)^{-1} d(x(t), t)\|, a_2 \|\nu(t)\|\} \Rightarrow \\ \dot{V} &\leq -a_3 \ell \|\varepsilon\|^2.\end{aligned}$$

As P is symmetric and definite positive, it turns out that $\underline{\lambda} \|\varepsilon\|^2 \leq V(\varepsilon) \leq \bar{\lambda} \|\varepsilon\|^2$ where $\underline{\lambda}$ and $\bar{\lambda}$ are respectively the smallest and the highest eigenvalue of P . By using these

bounds, the previous implication leads to conclude that, as long as

$$V \geq \bar{\lambda} \max\{a_1^2 \|D_n(\ell)^{-1} d(x(t), t)\|^2, a_2^2 \|\nu(t)\|\}$$

then $V(t) \leq \exp(-\frac{a_3 \ell}{\bar{\lambda}} t) V(0)$. By using again the bound on V in terms of $\underline{\lambda}$ and $\bar{\lambda}$, the following estimate on $\varepsilon(t)$ can be easily obtained

$$\|\varepsilon(t)\| \leq \max\{a_4 \exp(-a_5 \ell t) \|\varepsilon(0)\|, a_6 \|D_n(\ell)^{-1} d(x(t), t)\|, a_7 \|\nu(t)\|\}$$

where $a_4 = \sqrt{\bar{\lambda}/\underline{\lambda}}$, $a_5 = a_3/2\bar{\lambda}$, $a_6 = \sqrt{\bar{\lambda}/\underline{\lambda}} a_1$, $a_7 = \sqrt{\bar{\lambda}/\underline{\lambda}} a_2$. Now, using the fact that, for all $\ell > 1$, $\ell^{-(n-1)} \|\tilde{\xi}\| \leq \|\varepsilon\| \leq \|\tilde{\xi}\|$, the previous bound leads to the following estimate on $\tilde{\xi}(t)$

$$\begin{aligned}\|\tilde{\xi}(t)\| &\leq \max\{a_4 \ell^{n-1} \exp(-a_5 \ell t) \|\tilde{\xi}(0)\|, \\ &\quad a_6 \ell^{n-1} \|D_n(\ell)^{-1} d(x(t), t)\|, a_7 \ell^{n-1} \|\nu(t)\|\}\end{aligned}$$

by which the claim of the proposition immediately follows by bearing in mind the definition of $\Gamma(\ell)$, $\hat{\mathbf{x}}$, \mathbf{x} , and by noting that $\|\tilde{\xi}\| \leq \|\hat{\mathbf{x}} - \mathbf{x}\| \leq 2\|\tilde{\xi}\|$. \triangleleft

Remark 1 One might wonder if the fact of considering the observer (8) of order $2n-2$ leads to a value of ℓ^* , namely a lower bound of the high-gain parameter ℓ , that is higher than the one resulting from the use of a standard high-gain observer (4) of dimension n , by thus mitigating the benefit of a power of ℓ that grows up to the order 2 instead of n . It is readily seen that this is not the case. As a matter of fact, by going through the proof of Proposition 1, it follows that $\ell^* = 4\delta_1 \|P\|$ where δ_1 the Lipschitz constant of $\varphi(\cdot)$ in \mathcal{X} and $P > 0$ solution of the Lyapunov equation $PM + M^T P = -I$, with M defined in (7). Since P is positive definite and M is Hurwitz, it turns out that (see Theorem 2.4 in [8])

$$\|P\| \leq \bar{\lambda}(P) \leq \frac{1}{|\bar{\lambda}(M + M^T)|}$$

where $\bar{\lambda}(\cdot)$ denotes the maximum eigenvalue of the matrix in the argument. This shows that ℓ^* does not depend on the dimension of the matrix P (and thus on the observer's dimension) but only on the Lipschitz constant δ_1 of $\varphi(\cdot)$ and on the maximum eigenvalue of $M + M^T$.

III. ABOUT THE SENSITIVITY OF THE OBSERVER TO HIGH FREQUENCY NOISE IN THE LINEAR CASE

The trade-off between the speed of the state estimation and the sensitivity to measurement noise is a well-known fact in the observer theory. In this respect, high-gain observers tuned to obtain fast estimation dynamics are necessarily very sensitive to high-frequency noise. Bounds on the estimation error in presence of measurement noise for the standard high-gain observers have been studied, for instance, in [11] and [12], and different techniques have been developed in order to improve rejection, mainly based on gain adaptation (see, among others, [13], [14]).

In this section we compare the properties of the standard high-gain observer (4) and the proposed observer (8) with

respect to high-frequency measurement noise by specialising the analysis to linear systems.

In particular we consider systems of the form (2) with $\varphi(\cdot)$ a linear function of the form $\varphi(x) = \Phi x$ where Φ is a row vector of dimension n . Moreover, in this contest, we consider $d(t) \equiv 0$ and

$$\nu(t) = a_N \sin(\omega_N t + f_N) \quad (11)$$

where $a_N > 0$, $\omega_N > 0$, and f_N are constants. It is shown that the ratio between the asymptotic estimation error on the i -th state variable provided by the new observer (8) and the one provided by the standard observer (4) is a strictly decreasing polynomial function of the noise frequency for $i = 2, \dots, n$. In this regard the new observer has better asymptotic properties with respect to high-frequency noise as far as the state estimation variables are concerned (except for the first one). This is formalised in the next proposition.

Proposition 2 *Let $n \geq 1$. Let $K_n \in \mathbb{R}^n$, $K_i \in \mathbb{R}^2$, $i = 1, \dots, n-1$ and $\ell \in \mathbb{R}$ be fixed so that the error dynamics of the observers (4) and (8) are Hurwitz. Moreover, with $\rho \in \{1, 2, \dots, n\}$ denoting the position index of the first non zero coefficient of the vector Φ (and $\rho = n$ if Φ is the zero vector), let $r'_i \geq 1$ be the constants defined as*

$$r'_i = \min \{ i, (n-1), (\rho + n - i + 1) \} \quad i = 1, \dots, n.$$

There exist $\omega_N^ > 0$ and $\bar{c}_i > 0$ such that for all $\omega_N > \omega_N^*$, $a_N > 0$ and f_N we have*

$$\frac{\limsup_{t \rightarrow +\infty} |\hat{x}'_i(t) - x_i(t)|}{\limsup_{t \rightarrow +\infty} |\hat{x}_i(t) - x_i(t)|} \leq \bar{c}_i \omega_N^{-(r'_i-1)} \quad \forall i = 1, \dots, n.$$

Proof Consider system (2) and the standard high-gain observer (4). By letting $e = \ell D_n(\ell)^{-1} (\hat{x} - x)$, the e -dynamics read as

$$\dot{e} = \ell(A_n + K_n C_n) e + B_n \Phi \Theta_n(\ell) e + \ell K_n \nu(t) \quad (12)$$

where $\Theta_n(\ell) = \frac{1}{\ell^n} D_n(\ell)$ with $D_n(\ell)$ defined in (4). It is a linear system that is Hurwitz by the choices of K_n and ℓ . Similarly, consider system (2) and the new observer (8). With ε defined as in the proof of Proposition 1 we obtain system (10) compactly rewritten as

$$\dot{\varepsilon} = \ell M \varepsilon + B_{2n-2} \Phi \Theta_n(\ell) L_1 \varepsilon + \ell \bar{K}_1 \nu(t). \quad (13)$$

It is an Hurwitz system by the choices of K_i , $i = 1, \dots, n-1$, and ℓ . We consider now the n systems given by the dynamics (12) with input ν and with outputs

$$\hat{x}_i - x_i = \ell^{i-1} e_i, \quad i = 1, \dots, n,$$

and we denote by $F_i : \mathbb{R} \rightarrow \mathbb{C}$, $i = 1, \dots, n$, the harmonic transfer functions of these systems. A simple computation shows that these systems have relative degree $r_i = 1$, for all $i = 1, \dots, n$. Similarly, we consider the n systems given by the dynamics (13) with input ν and outputs

$$\begin{aligned} \hat{x}'_i - x_i &= \ell^{i-1} \varepsilon_{i,1} & i = 1, \dots, n-1, \\ \hat{x}'_n - x_n &= \ell^{n-1} \varepsilon_{n-1,2}, \end{aligned}$$

and we denote by $F'_i : \mathbb{R} \rightarrow \mathbb{C}$, $i = 1, \dots, n$, the harmonic transfer functions of these systems. Simple computations show that these systems have relative degree r'_i , for all $i = 1, \dots, n$, with r'_i defined in the statement of the proposition. By definition of harmonic transfer function and by the fact that systems (12) and (13) are Hurwitz, it turns out that

$$\begin{aligned} \limsup_{t \rightarrow +\infty} |\hat{x}_i(t) - x_i(t)| &= |F_i(\omega_N)| a_N \\ \limsup_{t \rightarrow +\infty} |\hat{x}'_i(t) - x_i(t)| &= |F'_i(\omega_N)| a_N \end{aligned}$$

for any $\omega_N \geq 0$ and $\forall i = 1, \dots, n$. Furthermore, by the fact that $F_i(\omega)$ and $F'_i(\omega)$ have, respectively, relative degrees $r_i = 1$ and r'_i , $i = \dots, n$, it turns out that there exist positive c_i , c'_i and $\omega_N^* > 0$ such that

$$\begin{aligned} |F_i(\omega)| &\geq c_i \omega^{-1} \\ |F'_i(\omega)| &\leq c'_i \omega^{-r'_i} \end{aligned} \quad \forall \omega \geq \omega_N^*,$$

by which the result immediately follows. \triangleleft

IV. EXAMPLE: OBSERVER FOR THE UNCERTAIN VAN DER POL OSCILLATOR

Let consider the uncertain Van der Pol oscillator

$$\ddot{z} = -\alpha^2 z + \beta(1 - z^2)\dot{z}, \quad y = z + \nu(t), \quad (14)$$

where $y \in \mathbb{R}$ is the measured output and α, β are uncertain constant parameters. We let $\mu = (\alpha^2, \beta)^\top$ and we assume that $\mu \in \mathcal{U}$, with \mathcal{U} a compact set of \mathbb{R}^2 not containing the origin. The state (z, \dot{z}) belongs to a compact invariant set $\mathcal{W} \subset \mathbb{R}^2$, which is the limit cycle of the Van Der Pol oscillator. We observe that \mathcal{W} depends on μ . Following [9], system (14) extended with $\dot{\mu} = 0$ can be immersed into a system in the canonical observability form (2) with $n = 5$. As a matter of fact, let $z_{[i,j]} := \text{col}(z^{(i)}, \dots, z^{(j)})$ be the vector of time derivatives of z , with $0 \leq i < j$, and let \mathcal{X} be the compact set of \mathbb{R}^5 such that $z_{[0,1]} \in \mathcal{W} \Rightarrow z_{[0,4]} \in \mathcal{X}$. Simple computations show that

$$z_{[2,4]} = \Upsilon(z_{[0,3]}) \mu, \quad z^{(5)} = \rho(z_{[0,4]}) \mu$$

where

$$\Upsilon(z_{[0,3]}) = \begin{pmatrix} -z & (1 - z^2)\dot{z} \\ -\dot{z} & \ddot{z} - 2y\dot{z}^2 - z^2\ddot{z} \\ -\ddot{z} & z^{(3)} - 2\dot{z}^3 - 6z\dot{z}\ddot{z} - z^2 z^{(3)} \end{pmatrix}$$

and

$$\rho(z_{[0,4]}) = (z^{(3)}, z^{(4)}(1 - z^2) - 12\dot{z}^2\ddot{z} - 6z\ddot{z}^2 - 8z\dot{z}z^{(3)})$$

with (see [10])

$$\text{rank} \Upsilon(z_{[0,3]}) = 2 \quad \text{for all } z_{[0,4]} \in \mathcal{X}.$$

Hence, by letting

$$\varphi(z_{[0,4]}) = \rho(z_{[0,4]}) \hat{\mu}(z_{[0,3]})$$

with

$$\hat{\mu}(z_{[0,4]}) = \Upsilon^\dagger(z_{[0,4]}) z_{[2,4]},$$

where $\Upsilon^\dagger(\cdot)$ is the left-inverse of $\Upsilon(\cdot)$, and letting $x = z_{[0,4]}$, it is immediately seen that system (14) and $\dot{\mu} = 0$ restricted to $\mathcal{W} \times \mathcal{U}$ is immersed into the system

$$\dot{x} = A_5 x + B_5 \varphi(x), \quad y = C_5 x + \nu(t), \quad (15)$$

where A_5, B_5, C_5 is a triplet in prime form of dimension 5.

By following the prescriptions of Section II, we implemented the proposed observer (8) as

$$\begin{aligned}\dot{\xi}_1 &= A \xi_1 + N \xi_2 + D_2(\ell_1) K_1 (y - C \xi_1) \\ &\vdots \\ \dot{\xi}_4 &= A \xi_4 + B \varphi_s(\hat{x}') + D_2(\ell_1) K_4 (B^\top \xi_3 - C \xi_4)\end{aligned}\quad (16)$$

where $\varphi_s(\cdot)$ is any locally Lipschitz bounded function that agrees with $\varphi(\cdot)$ on \mathcal{X} , $\xi_i = (\xi_{i1}, \xi_{i2})$ and the coefficients of K_i are

$$\begin{aligned}K_1 &= \begin{pmatrix} 0.6 \\ 0.3 \end{pmatrix}, & K_2 &= \begin{pmatrix} 0.6 \\ 0.111 \end{pmatrix}, \\ K_3 &= \begin{pmatrix} 0.6 \\ 0.0485 \end{pmatrix}, & K_4 &= \begin{pmatrix} 0.6 \\ 0.0178 \end{pmatrix},\end{aligned}$$

such that the roots of $\mathcal{P}_5(\lambda)$ are $-0.1, -0.2, -0.2, -0.3, -0.3, -0.4, -0.4, -0.5$. With the same notation of (8) we have $\xi = (\xi_1, \xi_2, \xi_3, \xi_4)$, $\hat{x}' = L_1 \xi$ and $\hat{x}'' = L_2 \xi$. In the simulations we fixed $\alpha = 1, \beta = 0.5$, gain $\ell_1 = 100$ and initial conditions $(z, \dot{z}) = (1, 0)$ for (14) and $\xi = 0$ for (16). Figures 1 and 2 show the error state estimate $|\hat{x}' - x|$ and $|\hat{x}'' - x|$ of the proposed observer (16) for the first two components (namely the estimation of the state of (14)) when there is not sensor noise. By following Section III we compared the observer (16) with a standard high-gain observer in presence of high-frequency sensor noise, numerically taken as $\nu(t) = 10^{-2} \sin(10^3 t)$. The high-gain observer has been implemented as

$$\dot{\hat{x}} = A_5 \hat{x} + B_5 \varphi_s(\hat{x}) + D(\ell_2) K_5 (y - C_5 \hat{x}) \quad (17)$$

where $K_5 = (1.5, 0.85, 0.225, 0.0274, 0.0012)^T$ so that the eigenvalues of $(A_5 + K_5 C_5)$ are $-0.1, -0.2, -0.3, -0.4, -0.5$, and $\ell_2 = 100$. Table 1 shows the normalized asymptotic error magnitudes of the proposed observer (16) and the standard high-gain observer (17), where the normalized asymptotic error for the i -th estimate is defined as $\|\hat{x}_i - x_i\|_a = \limsup_{t \rightarrow \infty} \|\hat{x}_i(t) - x_i(t)\|/a_N$. Although the result in Section III is given just for linear systems, the numerical results shown in the table show a remarkable improvement of the sensitivity of the new observer with respect to the standard high-gain observer.

| Standard High Gain Observer \hat{x} | Modified Observer $\hat{x}' = L_1 \xi$ | Modified Observer $\hat{x}'' = L_2 \xi$ |
|--|---|--|
| $\ \hat{x}_1 - x_1\ _a = 0.15$ | $\ \hat{x}'_1 - x_1\ _a = 0.06$ | $\ \hat{x}''_1 - x_1\ _a = 0.06$ |
| $\ \hat{x}_2 - x_2\ _a = 8$ | $\ \hat{x}'_2 - x_2\ _a = 0.2$ | $\ \hat{x}''_2 - x_2\ _a = 3$ |
| $\ \hat{x}_3 - x_3\ _a = 2 \cdot 10^2$ | $\ \hat{x}'_3 - x_3\ _a = 0.2$ | $\ \hat{x}''_3 - x_3\ _a = 3$ |
| $\ \hat{x}_4 - x_4\ _a = 2.5 \cdot 10^3$ | $\ \hat{x}'_4 - x_4\ _a = 0.1$ | $\ \hat{x}''_4 - x_4\ _a = 2$ |
| $\ \hat{x}_5 - x_5\ _a = 10^4$ | $\ \hat{x}'_5 - x_5\ _a = 0.3$ | $\ \hat{x}''_5 - x_5\ _a = 0.3$ |

Table 1: Normalized asymptotic errors in presence of noise.

V. CONCLUSIONS

We presented a new observer design based on high-gain techniques with a tunable state-estimate convergence speed. With respect to standard high-gain observers the state dimension is larger ($2n - 2$ instead of n) with a clear benefit in

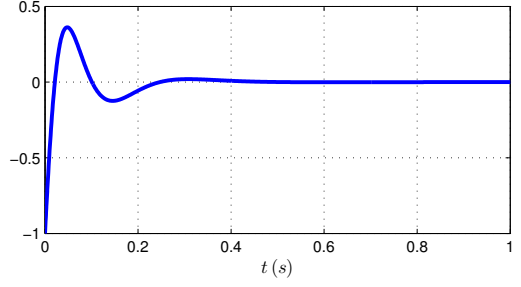


Fig. 1: Error state estimate $|\hat{x}''_1 - x_1|, |\hat{x}'_1 - x_1|$.

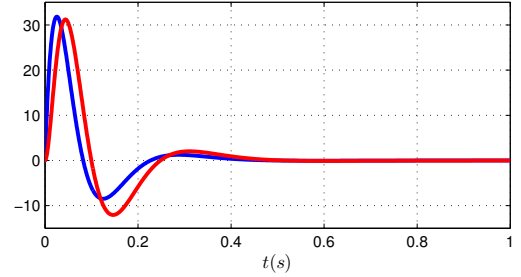


Fig. 2: Error state estimate $|\hat{x}''_2 - x_2|$ (blue line) and $|\hat{x}'_2 - x_2|$ (red line).

the observer implementation due to the power of the high-gain which is only 2 and not n . Moreover, when specialised to linear systems, we showed the benefit of the proposed observer with respect to the standard high-gain observer in terms of high-frequency noise rejection. Benefits that are clearly confirmed also for the nonlinear Van-der Pol example numerically simulated in the previous section. A complete characterisation of the sensitivity to sensor noise of the new observer is an interesting research topic that is now under investigation.

The peaking phenomenon due to wrong initial conditions and fast convergence that is typical of high-gain observers is not prevented by our proposed structure. However, other techniques to deal with peaking (such as saturations, time-varying gains [14], gradients techniques [18], and others) are available and can be adopted to improve the proposed observer structure. In this work we didn't consider the multi-output case. For the specific class of multi-output systems which are diffeomorphic to a block triangular form in which each block is associated to each output and it has a triangular dependence on the states of that subsystem (see [19]), the proposed structure can be simply applied block-wise to obtain a high-gain observer. Apart this case, a complete extension to the multi-output case is not immediate and under investigation.

Acknowledgment. We wish to thank Laurent Praly for suggesting the design procedure presented in Appendix A.

APPENDIX

A. Procedure to assign the eigenvalues of M

Consider the matrices $M_i \in \mathbb{R}^{2i-2} \times \mathbb{R}^{2i-2}$ recursively defined as

$$M_1 = E_1, \quad M_i = \begin{pmatrix} M_{i-1} & \bar{N}_i \\ \bar{Q}_i & E_i \end{pmatrix}, \quad i = 2, \dots, n-1$$

where $\bar{N}_i = \text{col}(0_{2(i-2) \times 2}, N)$, $\bar{Q}_i = (0_{2(i-2) \times 2}, Q_i)$ and E_i , $i = 1, \dots, n-1$, Q_i , $i = 2, \dots, n-1$, and N are defined as in the definition of M . Note that $M = M_{n-1}$ and, by letting $K_i = (k_{i1} \ k_{i2})^T$, note that Q_i and E_i depend on K_i , while M_i depends on K_1, \dots, K_i . We let $\mathcal{P}_{M_i}(\lambda) = \lambda^{2i} + m_1^i \lambda^{2i-1} + \dots + m_{2i-1}^i \lambda + m_{2i}^i$ and $\mathcal{P}_{M_{i-1}}(\lambda) = \lambda^{2i-2} + m_1^{i-1} \lambda^{2i-3} + \dots + m_{2i-3}^{i-1} \lambda + m_{2i-2}^{i-1}$ the characteristic polynomials of M_i and M_{i-1} , and we use the notation $m_{[1,j]}^i = \text{col}(m_1^i, \dots, m_j^i) \in \mathbb{R}^j$, $m_{[1,k]}^{i-1} = \text{col}(m_1^{i-1}, \dots, m_k^{i-1}) \in \mathbb{R}^k$ for some $j \leq 2i$ and $k \leq 2i-2$.

The characteristic polynomial of $\mathcal{P}_{M_i}(\lambda)$ is computed as

$$\begin{aligned} \mathcal{P}_{M_i}(\lambda) &= \lambda(\lambda + k_{i1})\mathcal{P}_{M_{i-1}}(\lambda) \\ &\quad + k_{i2} [\mathcal{P}_{M_{i-1}}(\lambda) - \lambda(\lambda + k_{(i-1)1})\mathcal{P}_{M_{i-2}}(\lambda)]. \end{aligned}$$

Hence, simple, although lengthy, computations show that the coefficients $m_{[1,2i]}^i$ of $\mathcal{P}_{M_i}(\lambda)$ and $m_{[1,2i-2]}^{i-1}$ of $\mathcal{P}_{M_{i-1}}(\lambda)$ are related as follow

$$\begin{aligned} m_{[1,2i-2]}^i &= (I_{2i-2} + k_{i1}F)m_{[1,2i-2]}^{i-1} + k_{i1}v_1 \\ m_{2i-1}^i &= k_{i1}m_{2i-2}^{i-1} \\ m_{2i}^i &= k_{i2}m_{2i-2}^{i-1} \end{aligned} \quad (18)$$

where $v_1 \in \mathbb{R}^{2i-2}$ is the zero vector with a 1 in the first position, and $F \in \mathbb{R}^{(2i-2) \times (2i-2)}$ is the zero matrix with the identity matrix I_{2i-3} in the lower left block. Note that $(I_{2i-2} + k_{i1}F)$ is invertible for all k_{i1} . Hence, from the first of (18), one obtains

$$m_{[1,2i-2]}^{i-1} = \Lambda(m_{[1,2i-2]}^i, k_{i1})$$

where

$$\Lambda(m_{[1,2i-2]}^i, k_{i1}) = (I_{2i-2} + k_{i1}F)^{-1} (m_{[1,2i-2]}^i - k_{i1}v_1),$$

which, embedded in the second and in the third of (18), yield the relations

$$\sigma_1(m_{[1,2i-1]}^i, k_{i1}) = 0, \quad k_{i2} = \sigma_2(m_{[1,2i]}^i, k_{i1})$$

where

$$\begin{aligned} \sigma_1(m_{[1,2i-1]}^i, k_{i1}) &= k_{i1}v_2^T \Lambda(m_{[1,2i-2]}^i, k_{i1}) - m_{2i-1}^i \\ \sigma_2(m_{[1,2i]}^i, k_{i1}) &= \frac{m_{2i}^i}{v_2^T \Lambda(m_{[1,2i-2]}^i, k_{i1})} \end{aligned}$$

in which $v_2 \in \mathbb{R}^{2i-2}$ is the zero vector with a 1 in the last position. We observe that $\sigma_1(\cdot, \cdot)$ is a polynomial in k_{i1} of odd order $2i-1$. As a consequence, for any $m_{[1,2i-1]}^i$ there always exists at least one real k_{i1} fulfilling $\sigma_1(m_{[1,2i-1]}^i, k_{i1}) = 0$.

The previous results can be used to set up a "basic assignment algorithm" that is then used iteratively to solve the eigenvalues assignment of the matrix M .

Basic assignment algorithm. Let $\bar{\mathcal{P}}_i(\lambda) = \lambda^{2i} + \bar{m}_1^i \lambda^{2i-1} +$

$\dots + \bar{m}_{2i}^i$ be an arbitrary polynomial. Then, there exist a real $\bar{K}_i = (\bar{k}_{i1}, \bar{k}_{i2})^T$ and a polynomial $\bar{\mathcal{P}}_{i-1}(\lambda) = \lambda^{2i-1} + \bar{m}_1^{i-1} \lambda^{2i-2} + \dots + \bar{m}_{2i-2}^{i-1}$ such that

$$\begin{aligned} K_i &= \bar{K}_i \\ \mathcal{P}_{M_{i-1}} &= \bar{\mathcal{P}}_{i-1}(\lambda) \end{aligned} \quad \Rightarrow \quad \mathcal{P}_{M_i}(\lambda) = \bar{\mathcal{P}}_i(\lambda).$$

As a matter of fact, by letting $\bar{m}_{[1,2i-1]}^i$ the coefficients of $\bar{\mathcal{P}}_i(\lambda)$, it is possible to take \bar{k}_{i1} as a real solution of $\sigma_1(\bar{m}_{[1,2i-1]}^i, \bar{k}_{i1}) = 0$, $\bar{k}_{i2} = \sigma_2(\bar{m}_{[1,2i]}^i, \bar{k}_{i1})$, and to take the coefficients $\bar{m}_{[1,2i-2]}^{i-1}$ of the polynomial $\bar{\mathcal{P}}_{i-1}(\lambda)$ as $\bar{m}_{[1,2i-2]}^{i-1} = \Lambda(\bar{m}_{[1,2i-2]}^i, \bar{k}_{i1})$. \triangleleft

With the previous algorithm in hand, the design of K_1, \dots, K_{n-1} to assign an arbitrary characteristic polynomial to M , can be then immediately done by the following steps:

- 1) With $\bar{\mathcal{P}}_{n-1}(\lambda)$ the desired characteristic polynomial of M , compute $(\bar{K}_{n-1}, \bar{\mathcal{P}}_{n-2}(\lambda))$ by running the basic assignment algorithm with $i = n-1$.
- 2) Compute iteratively $(\bar{K}_i, \bar{\mathcal{P}}_{i-1}(\lambda))$ by running the basic assignment algorithm for $i = n-2, \dots, 2$.
- 3) Compute $\bar{K}_1 = (\bar{k}_{11}, \bar{k}_{12})^T$ so that $\lambda^2 + k_{11}\lambda + k_{12} = \bar{\mathcal{P}}_1(\lambda)$.

REFERENCES

- [1] J. P. Gauthier, G. Bornard, "Observability for any $u(t)$ of a class of nonlinear systems", IEEE Transactions on Automatic Control, Vol. AC-26, No. 4, August 1981.
- [2] J.P. Gauthier and I. Kupka, "Deterministic observation theory and applications", Cambridge University Press, 2004.
- [3] G. Bornard and H. Hammouri, "A high gain observer for a class of uniformly observable systems", IEEE, 30th Conference on Decision and Control, December 1991.
- [4] G. Besançon, *Nonlinear observers and applications*, Springer, 2007.
- [5] A. Teel and L. Praly, "Global stabilizability and observability imply semi-global stabilizability by output feedback", Systems and Control Letters, 22, pp 313-325, 1994.
- [6] A. N. Atassi and H. K. Khalil, "A separation principle for the stabilization of a class of nonlinear systems", IEEE Transactions on Automatic Control, 44, pp 1672-1687, 1999.
- [7] H. K. Khalil and L. Praly, "High-gain observers in nonlinear feedback control", Robust and Nonlinear Control, 24, pp 993-1015, 2014.
- [8] M. T. J. Gajić, *The Lyapunov matrix equation in system stability and control*, Academic Press, 1995.
- [9] D. Astolfi, L. Praly and L. Marconi, "A note on observability canonical forms for nonlinear Systems", IFAC, 9th Symposium on Nonlinear Control Systems, pp 436-438, September 2013.
- [10] F. Forte, A. Isidori, L. Marconi, "Robust design of internal models by nonlinear regression", IFAC, 9th Symposium on Nonlinear Control Systems, pp 283-288, September 2013.
- [11] L. K. Vasiljevic and H. K. Khalil, "Differentiation with high-gain observers the presence of measurement noise", IEEE 45th Conference on Decision and Control, pp 4717-4722, December 2006.
- [12] A. A. Ball and H. K. Khalil, "High-gain observer tracking performance in the presence of measurement noise", American Control Conference, pp 4626-4627, 2009.
- [13] J. H. Ahrens and H. K. Khalil, "High-gain observer in the presence of measurement noise: a switched-gain approach", Automatica vol. 45, pp 936-943, 2009.
- [14] R. G. Sanfelice and L. Praly, "On the performance of high-gain observers with gain adaptation under measurement noise", Automatica Vol. 47, pp 2165-2176, 2011.
- [15] H. Hammouri, G. Bornard and K. Busawon, "High gain observer for structured multi-input multi-output nonlinear systems", IEEE Transaction on Automatic Control vol. 55, no. 4, pp 987-992, April 2010.
- [16] F. Deza, E. Busvelle, J.P. Gauthier, D. Rakotopara, "High gain estimation for nonlinear systems", Systems & Control Letters vol. 18, pp 295-299, April 1992.

- [17] M. Maggiore, K.M. Passino, “A separation principle for a class of non uniformly completely observable systems”, IEEE Transactions on Automatic Control, vol. 48, no. 7, pp 1122-1133, July 2003.
- [18] D. Astolfi and L. Praly, “Output feedback stabilization for SISO nonlinear systems with an observer in the original coordinates”, IEEE 52nd Conference on Decision and Control, pp. 5927-5932, December 2013.
- [19] H. Shim, Y. I. Son, J. H. Seo “Semi-global observer for multi-output nonlinear systems”, System & Control Letters, vol. 42, pp 233-244, 2001.