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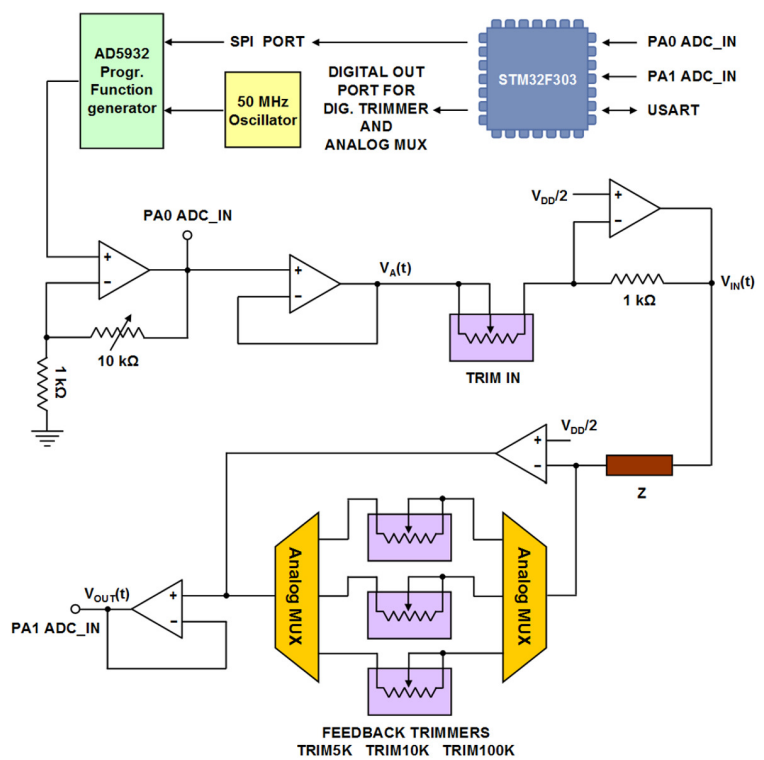
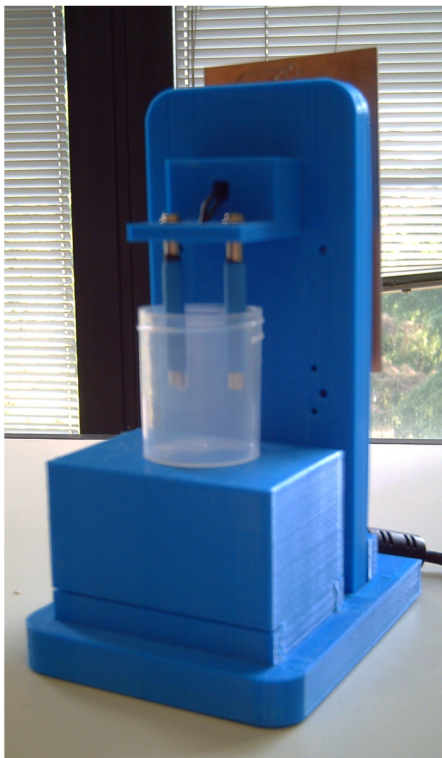
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**Electrical Impedance Spectroscopy
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abstract info.

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Electrical Impedance Spectroscopy (EIS) characterization of saline solutions with a low-cost portable measurement system

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ABSTRACT

Electrical Impedance Spectroscopy (EIS), a powerful technique used for wide range of applications, is usually carried out by means of benchtop instrumentation (LCR meters and impedance analyzers), not suited for in-the-field measurements performed outside a laboratory.

In this paper a new portable electronic system for EIS on liquid and semi-liquid media is presented that is capable to produce an electrical fingerprint of the sample under investigation. The proposed system was used for the characterization of four different saline solutions (NaCl, Na₂CO₃, K₂HPO₄ and CuSO₄). A multi-frequency approach, based on the measurement of maximum value of the impedance imaginary component and its corresponding frequency, was tested for the first time to discriminate different saline solutions. The results show that the proposed method is capable to discriminate the different solutions and to measure the concentration ($R^2 = 0.9965$) independently of the type of saline solution.

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1. Introduction

Electrical Impedance Spectroscopy (EIS) is a technique used in a wide range of applications [1]: from detection of bacterial concentration [2-8] to the analysis of human body composition [9-12], the characterization and quality assessment of different foods [13-18] as well as the investigation of corrosion of metals [19-21] and coated metal surfaces [22-24]. Other applications of EIS include the characterization of state-of-charge and state-of-health of batteries and fuel cells [25-28], the automated titration for different compounds [29] and the detection of cancer cells [30-34].

EIS was used for the first time in 1894 by W. Nerst to measure the dielectric constant of

different types of electrolytes [35]. However, it was only in the 1980s that the use of this technique really ramped up, thanks to computer-controlled laboratory instrumentation that allowed much easier implementations. A conference dedicated to research and applications of EIS is held every three years and the number of publications on the subject doubles every four or five years, with about 1200 papers published in 2006 [36].

Today EIS investigations are usually carried out using dedicated benchtop instruments, such as LCR meters and Impedance Analyzers, that provide: a) accurate measurements on a wide range of frequencies; b) a variety of operative modes (i.e. 2, 3 or 4 electrodes measurements); c) interfaces toward computers to transfer the acquired data for filing and further processing. Moreover, many commercial instruments are sold with proprietary software for EIS data analysis, where the acquired data are fitted by means of a user-selectable equivalent circuit made of resistors, capacitors and inductors, as well as of non-linear empirical elements such as Constant Phase Elements (CPE) [37]

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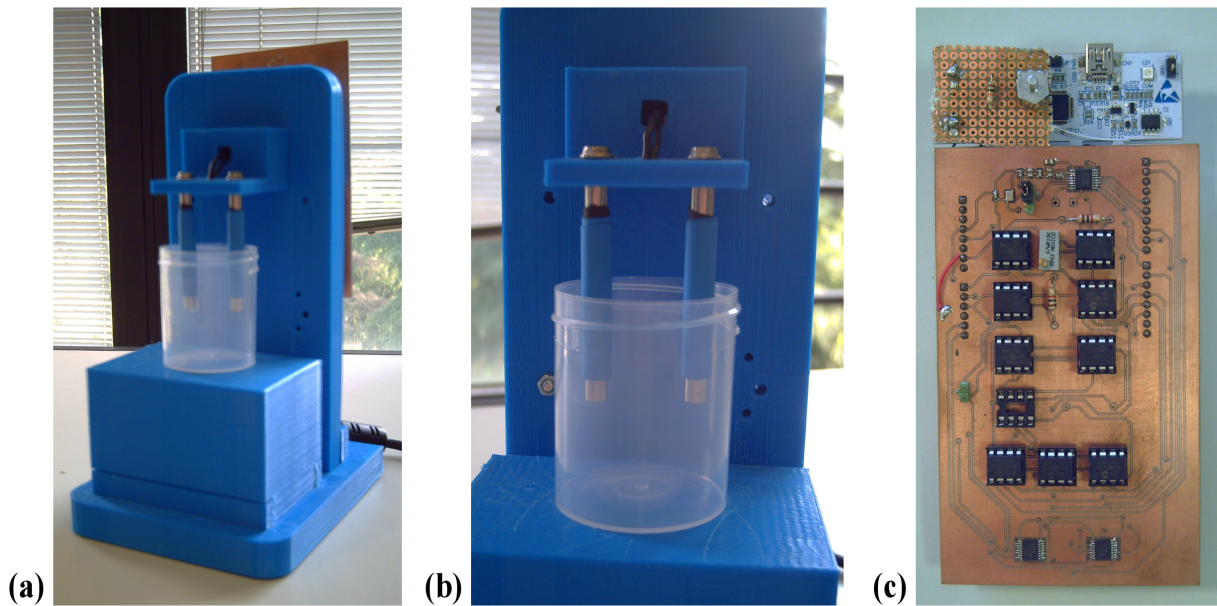


Fig. 1. Picture of the portable electronic system for Impedance Spectroscopy of liquid and semi-liquid media (a) and (b); electronic board (c).

and Warburg impedances [38]. However such scientific instruments are not suitable for in-situ and on-line analysis outside a laboratory.

On the other hand there are many applications requiring measurements outside a laboratory, such as for example: detection of milk adulteration [39] and clinical mastitis in cow's milk [40]; the analysis and characterization of fruits [41]; monitoring of alcoholic fermentation in wine [42]; detection of artificial chemical additives in liquid food products [43]; measurements of phenolic compounds' concentration in water [44]; analysis of petroleum products and characterization of water-in-oil emulsions [45]; detection of the underpaint corrosion of metallic surfaces [46].

Thus, in recent years, much effort has been devoted to the development of portable impedance analyzers to be used for in-situ EIS applications. This has become even simpler with the introduction on the market of the IC AD5933 (Analog Devices) [47], a low-cost on-chip impedance analyzer that can perform measurements in the impedance range $1k\Omega - 10 M\Omega$ (that can be extended to $100 \Omega - 10 M\Omega$ with additional hardware) and frequency range $1 kHz - 100 kHz$. Portable [48-50], wearable [51] as well as Smartphone based [52, 53] impedance analyzers based on AD5933 have been proposed in literature.

In this paper a new low-cost (about 100 USD) portable instrument for EIS analysis of liquid and semi-liquid media is presented. Such a system, built using a 3D printer, features a microcontroller based board for measurements and data transfer to a laptop PC. Differently from other portable instruments for EIS measurements discussed in literature, the proposed instruments

does not feature the IC AD5933 but uses an external waveform generator IC (AD5932) to generate the test signal and acquires the resulting signals using the analog-to-digital converters embedded in the microcontroller. This results in a reduction of the overall cost for the system as well as in the possibility to select different values for the amplitude of the AC test signal. Since it can be used by anyone and everywhere, it is suitable for in-field measurements outside a laboratory.

The proposed system has been tested to measure the concentration of different saline solutions, important particularly (but not exclusively) in the environmental monitoring, in the medical and biological fields since they are widely used as a vehicle to inject pharmaceuticals or as a supporting medium for biological and microbiological applications [54]. The obtained experimental results have shown that our system is able to estimate the solution concentration with good accuracy ($R^2 > 0.99$) independently of the type of saline solution.

2. Experimental approach

The proposed system ($10 \times 12 \times 18$ cm), shown in Fig. 1, is composed of a Polylactic Acid (PLA) structure realized with a 3D printer (MakerBot Replicator Z18) and an ad-hoc designed electronic board devoted to EIS measurements on samples hosted in disposable polypropylene containers. The system features a couple of cylinder shaped (radius 2.25 mm, length 5 mm, spaced 22 mm) stainless steel electrodes in direct contact with the Sample Under Test (SUT), hosted inside a 40 ml container. A temperature sensor (B57045K produced by TDK) is integrated on the electronic board to

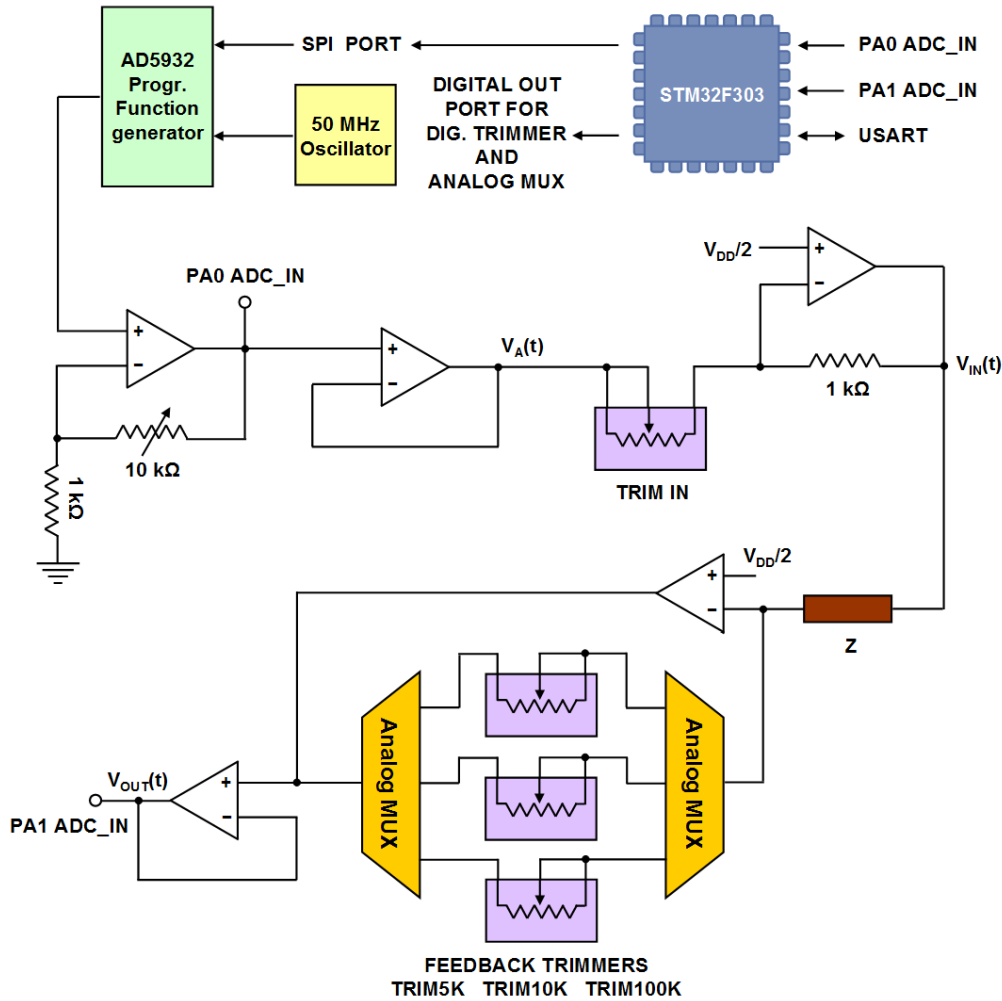


Fig. 2. Schematic of the designed electronic board.

measure the SUT temperature, assumed to be in thermal equilibrium with the environment.

2.1. EIS basics

The proposed system performs EIS measurements in potentiostatic mode. Thus, a sine-wave voltage test signal

$$V_{in}(t) = V_{M,in} \cdot \sin(\omega t) \quad (1)$$

is applied between the two electrodes and the current through them

$$I_{in}(t) = I_{M,in} \cdot \sin(\omega t + \varphi) \quad (2)$$

is measured. In the above equations, $V_{M,in}$ and $I_{M,in}$ are the voltage and current amplitude, respectively; $\omega = 2\pi f$ is the angular frequency (with f the frequency of the test signal); φ is the phase difference between $I_{in}(t)$ and $V_{in}(t)$. From equations (1) and (2) the complex impedance Z can be calculated as:

$$Z(j\omega) = \frac{V_{in}(j\omega)}{I_{in}(j\omega)} = \frac{V_{M,in}}{I_{M,in}} \cdot e^{-j\varphi} = \text{Re}(Z) + j \cdot \text{Im}(Z) \quad (3)$$

where $V_{in}(j\omega)$ and $I_{in}(j\omega)$ are the Steinmetz transforms of $V_{in}(t)$ and $I_{in}(t)$.

A sequence of test signals with increasing frequencies is applied to the SUT and the impedance, as well as its real and imaginary components $\text{Re}(Z)$ and $\text{Im}(Z)$, are calculated for all applied frequencies.

The acquired spectrum is normally displayed using either the Bode plots, where $\text{Re}(Z)$ and $\text{Im}(Z)$, or alternatively the corresponding admittance components $\text{Re}(Y)$ and $\text{Im}(Y)$, are plotted vs. frequency (in logarithmic scale), or the Nyquist plot, where the $\text{Im}(Z)$ or $\text{Im}(Y)$ is plotted vs. the corresponding $\text{Re}(Z)$ or $\text{Re}(Y)$ for all the investigated frequencies.

Usually, the measured data are fitted with an equivalent electrical circuit and the circuit parameters are used to investigate the correlation with the SUT parameters of interest. In our case, the data fitting procedure is made by means of the software tool Multiple Electrochemical Impedance Spectra Parameterization (MEISP) v 3.0 by Kumho Chemical Laboratories.

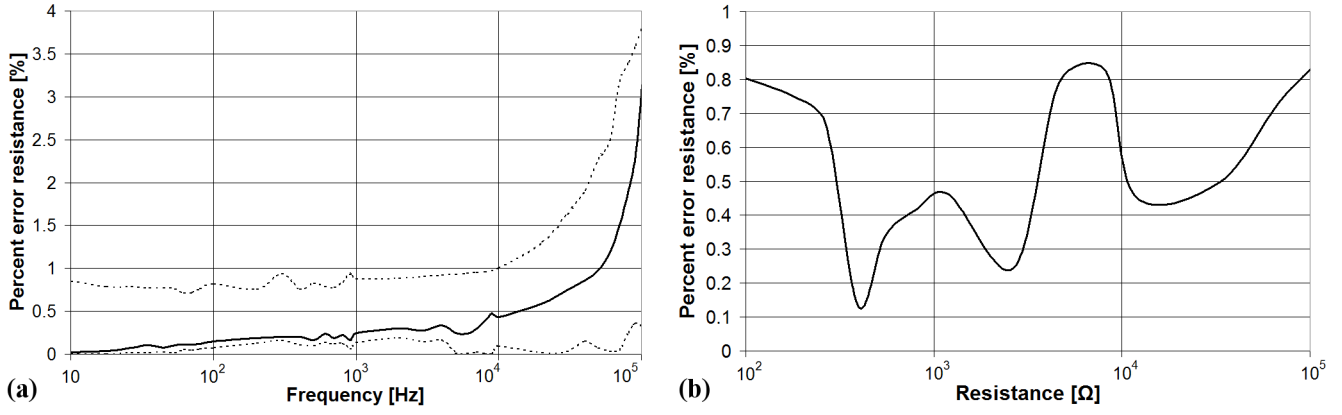


Fig. 3. Percent error on measured resistance (full line curve for a resistance value of 2680 Ω , dashed lines for the minimum and maximum values among all tested resistances) as function of frequency (a) and resistance value for frequency 1 kHz (b).

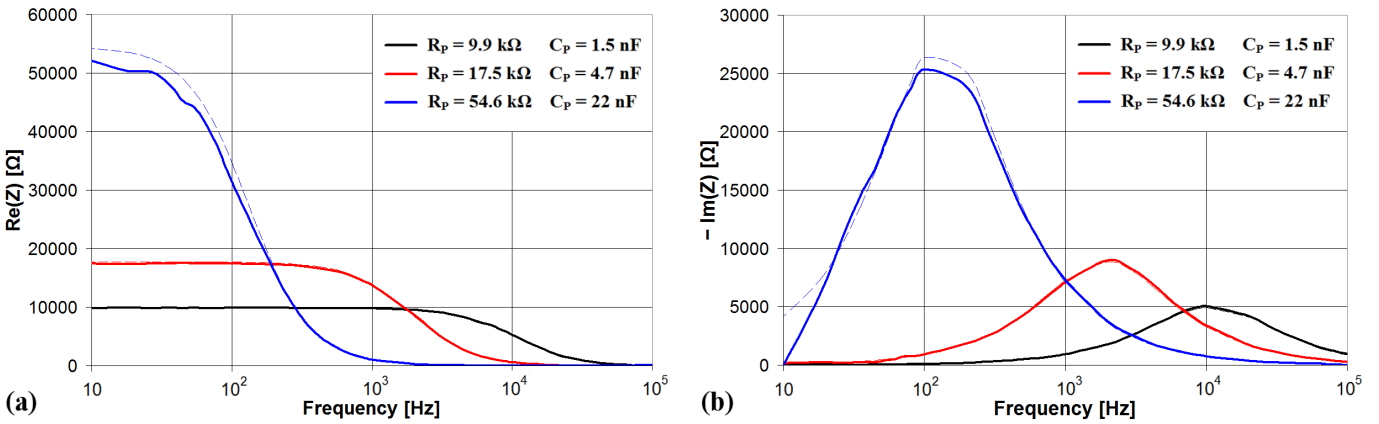


Fig. 4. Measured and theoretical values for $\text{Re}(Z)$ and $-\text{Im}(Z)$ in the case of impedances built from discrete resistances and capacitances.

2.2. Electronic board for EIS measurement

The electronic board of our system, designed ad-hoc using the software Kicad [55], is interfaced with a Nucleo STM32 development board (STM32F303 microcontroller by ST Microelectronics), that features a built-in UART-to-USB controller for easy data transfer to a laptop PC for filing and further processing. Fig. 1 (c) shows a picture of the designed board.

The schematic of the board is shown in Fig. 2. A sine-wave voltage signal (290 mV AC amplitude with 346 mV DC average value) is generated using an AD5932 programmable function generator that is controlled by the microcontroller through a SPI interface. This signal is then amplified with a non-inverting amplifier to achieve an AC amplitude of 1.38 V, with average DC value $V_{DD}/2 = 1.65$ V. Finally, a further inverting amplifier generates the test signal $V_{in}(t)$ (i.e. a sinusoidal voltage with DC value 1.65 V and amplitude controlled by the digital potentiometer TRIM_IN), that is applied to the electrodes in direct contact with the SUT.

The current through the electrodes is then converted to a proportional voltage $V_{out}(t)$ by

means of a current-to-voltage (I/V) converter, whose feedback digital potentiometers (TRIM_5K, TRIM_10K and TRIM_100K) can be selected by using two analog multiplexers to allow measurements over a wide range of impedances.

All the digital potentiometers and analog multiplexers are controlled by the microcontroller with digital signals. If

$$V_{in}(t) = \frac{V_{DD}}{2} + V_{M,in} \cdot \sin(\omega t) \quad (4)$$

the output signal

$$V_{out}(t) = \frac{V_{DD}}{2} + V_{M,out} \cdot \sin(\omega t + \varphi) \quad (5)$$

where V_{DD} is the microcontroller supply voltage (3.3 V) and the parameters $V_{M,in}$, $V_{M,out}$, φ and ω are calculated from $V_{in}(t)$ and $V_{out}(t)$ using the analog-to-digital converters (ADCs) integrated in the microcontroller and using a non-iterative algorithm based on the minimization of the mean square error [56]. The complex impedance can thus be calculated with equation 3 (where $I_{M,in}$ is

Table 1. Estimated values of resistance and capacitance from the fitting of the measured impedance spectra.

	$V_{M,in} = 1 \text{ V}$		$V_{M,in} = 500 \text{ mV}$		$V_{M,in} = 100 \text{ mV}$	
	R (k Ω)	C (nF)	R (k Ω)	C (nF)	R (k Ω)	C (nF)
$R_p = 9.9 \text{ k}\Omega$ $C_p = 1.5 \text{ nF}$	9.94	1.56	9.91	1.57	9.88	1.92
$R_p = 17.7 \text{ k}\Omega$ $C_p = 4.7 \text{ nF}$	17.4	4.64	17.5	4.67	17.1	5.14
$R_p = 54.6 \text{ k}\Omega$ $C_p = 22 \text{ nF}$	51.3	21.7	51.9	22.4	50.1	24.7
$R_s = 2.18 \text{ k}\Omega$ $C_s = 306 \text{ nF}$	2.16	309	2.17	309	2.16	314
$R_s = 21.8 \text{ k}\Omega$ $C_s = 306 \text{ nF}$	21.7	308	21.7	309	21.5	307
$R_s = 56.3 \text{ k}\Omega$ $C_s = 306 \text{ nF}$	56.3	309	55.9	309	55.9	289

determined from $V_{M,out}$ and the value of the I/V converter feedback resistance).

The system measures $\text{Re}(Z)$ and $\text{Im}(Z)$ spectra by testing the SUT with 37 different frequencies in the range 10 Hz - 100 kHz. The test signal amplitude can be selected by the user among three different values (100 mV, 500 mV and 1V) by setting the proper value of the digital potentiometer TRIM_IN.

A software interface, developed in LabVIEW (National Instruments), allows the user to set the measurement parameters, display the results as Bode and Nyquist plots as well as save the data on the hard disk for further analysis.

3. Results and discussion

The system discussed in Section 2 has been first tested with known impedances (realized by connecting discrete resistors and capacitors) and then using lab-prepared saline solutions.

3.1. Measurements on test impedances

A set of 14 discrete resistances (81.8, 178.3, 267, 393, 547, 811, 1200, 2680, 4630, 8205, 11875, 33200, 68750 and 120300 Ω) has been measured with the designed electronic board and the results show good accuracy compared with the values previously obtained with the LCR meter Agilent E4980A (errors < 1% for frequencies < 10 kHz).

Fig. 3 shows the percent error vs. frequency for a resistance value of 2680 Ω (a) and vs the resistance value for a test signal of frequency 1 kHz (b). In both cases the signal amplitude is set to 500 mV. In the case of Fig. 3 (a) the full line represents the percent error in the case of a resistance value of 2680 Ω , while the dashed lines represents the minimum and maximum percent error for the full set of tested resistances: as can be seen the resistance error is below 1% for all tested resistances when $f < 10$ kHz, while it increases for higher frequencies up to about 3.5% for 100 kHz.

Then the following discrete impedances have been prepared and tested with the electronic board: 1) 9.9 k Ω in parallel with 1.5 nF, 2) 17.5 k Ω in parallel with 4.7 nF, 3) 54.6 k Ω in parallel

with 22 nF, 4) 2.18 k Ω in series with 306 nF, 5) 21.8 k Ω in series with 306 nF, 6) 56.3 k Ω in series with 306 nF. Fig. 4 shows the Bode plots for $\text{Re}(Z)$ and $-\text{Im}(Z)$ in three different cases, with full lines representing measured values with the proposed system and dashed lines the expected results based on the values of resistance and capacitance measured with the commercial LCR meter.

The spectra measured for the six test impedances have been processed with the MEISP software (using the corresponding electrical circuit) to extract the value of resistance and capacitance. The results are presented in Table 1 for all test impedances and three values of signal amplitude (100 mV, 500 mV, 1V). As can be seen, the estimated values are accurate (error lower than 1%) in most cases with higher error in the case of high impedances and small test signal (100 mV).

Overall these tests indicate that the designed board can be reliably used for EIS measurements in the frequency range 10 Hz - 100 kHz on a wide range of impedances (80 Ω to 120 k Ω).

3.2. Measurements on saline solutions

Four different saline solutions have been prepared: sodium chloride (NaCl) 2M, sodium carbonate (Na_2CO_3) 2M, potassium hydrogen phosphate (K_2HPO_4) 2M, copper(II) sulfate (CuSO_4) 1M. From each solution, different dilutions in distilled water have been created to obtain different solute molecule concentrations (hereafter simply concentration). All measurements have been carried out in triplicate at the environmental temperature $T = 23.5$ $^\circ\text{C}$ and with a voltage amplitude of 100 mV. The aim of this work is to show the feasibility of saline solution discrimination by analysis of the EIS characteristics on a wide range of frequencies and thus measurements have been carried out at constant temperature. Since the electrical conductivity of electrolytic solutions is known to increase with temperature [57], a temperature sensor has been added to the system (as discussed in section 2). The measured temperature can thus be used to compensate the measurements for temperature variations according to some pre-calculated calibration equations.

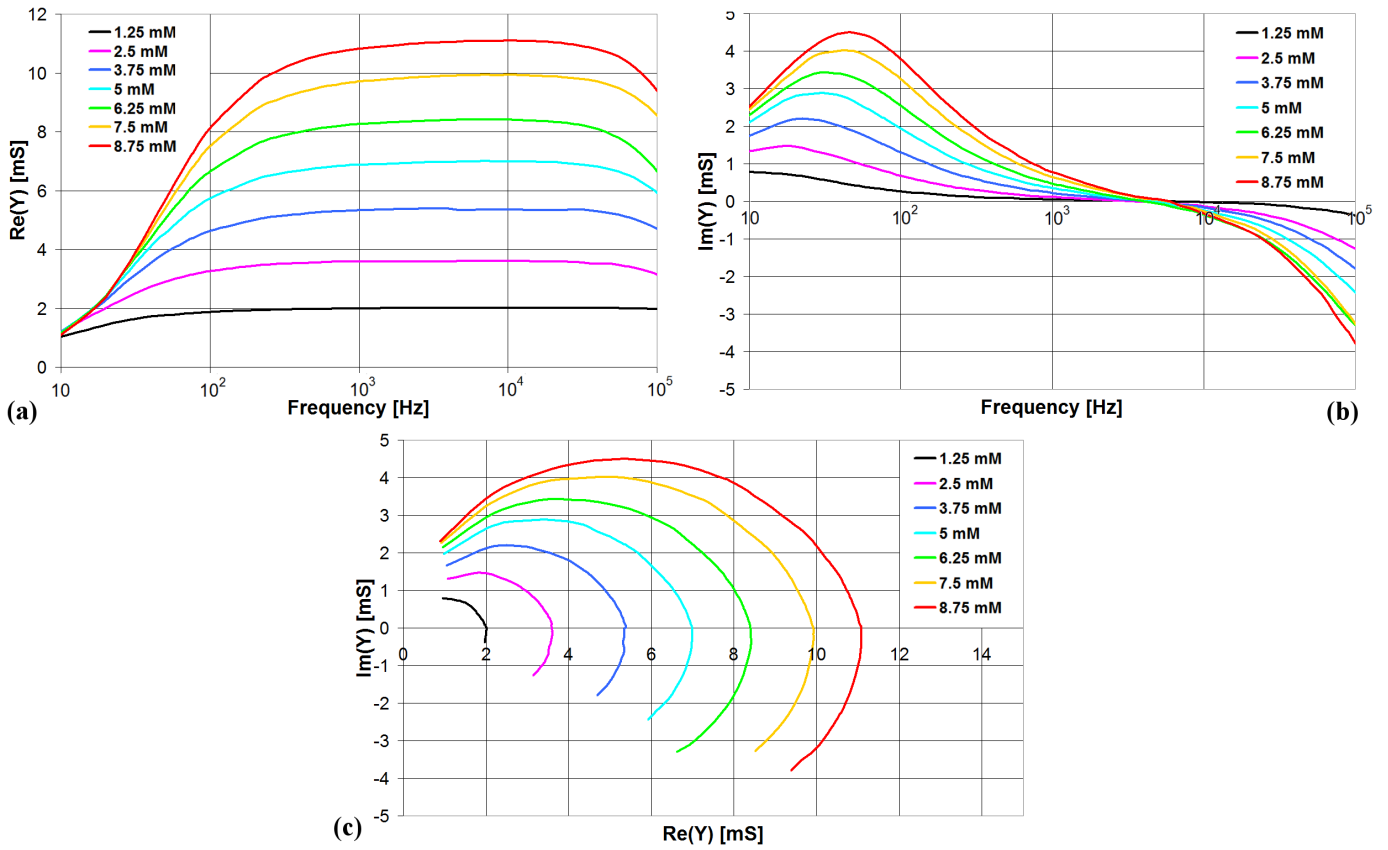


Fig. 5. Bode plots (a) and (b) and Nyquist plot (c) in the case of saline solutions of Na_2CO_3 of different molar concentrations.

The Bode and Nyquist plots for different concentrations of the saline solution Na_2CO_3 are presented in Fig. 5. As can be seen in Fig. 5 (b), the imaginary component of the admittance Y features a typical inductive behavior ($\text{Im}(Y) < 0$) at high frequencies (> 10 kHz): this is possibly caused by parasitic effects due to mutual inductance of wires and metal parts (electrodes). A proposed solution to decrease this effect is to use a 4-electrodes measurement set-up [58]. However, this allow the measurement of only the solution conductance and not the interfacial capacitance, thus preventing the correct classification of saline solutions as discussed later. Therefore, the corresponding high frequency (> 10 kHz) data are discarded in the following analysis.

In the frequency range 10 Hz - 10 kHz the Nyquist plot (Fig. 5 (c)) is represented by a semi-circle with its center below the x-axis, and this behavior can be modeled with an equivalent electrical circuit composed of an electrical conductance G_m in series with a constant phase element (CPE) used to model a non-ideal capacitive interface. The resulting impedance is then:

$$Z = \frac{1}{G_m} + \frac{1}{Q(j\omega)^\alpha} = \frac{1}{G_m} + \frac{\cos\left(\frac{\alpha\pi}{2}\right)}{Q\omega^\alpha} - j \frac{\sin\left(\frac{\alpha\pi}{2}\right)}{Q\omega^\alpha} \quad (6)$$

where Q is the capacitance of the CPE and α is a parameter accounting for the non-ideality of Q (if $\alpha = 1$ the CPE becomes an ideal capacitance).

The real and imaginary components of Y can be modeled as:

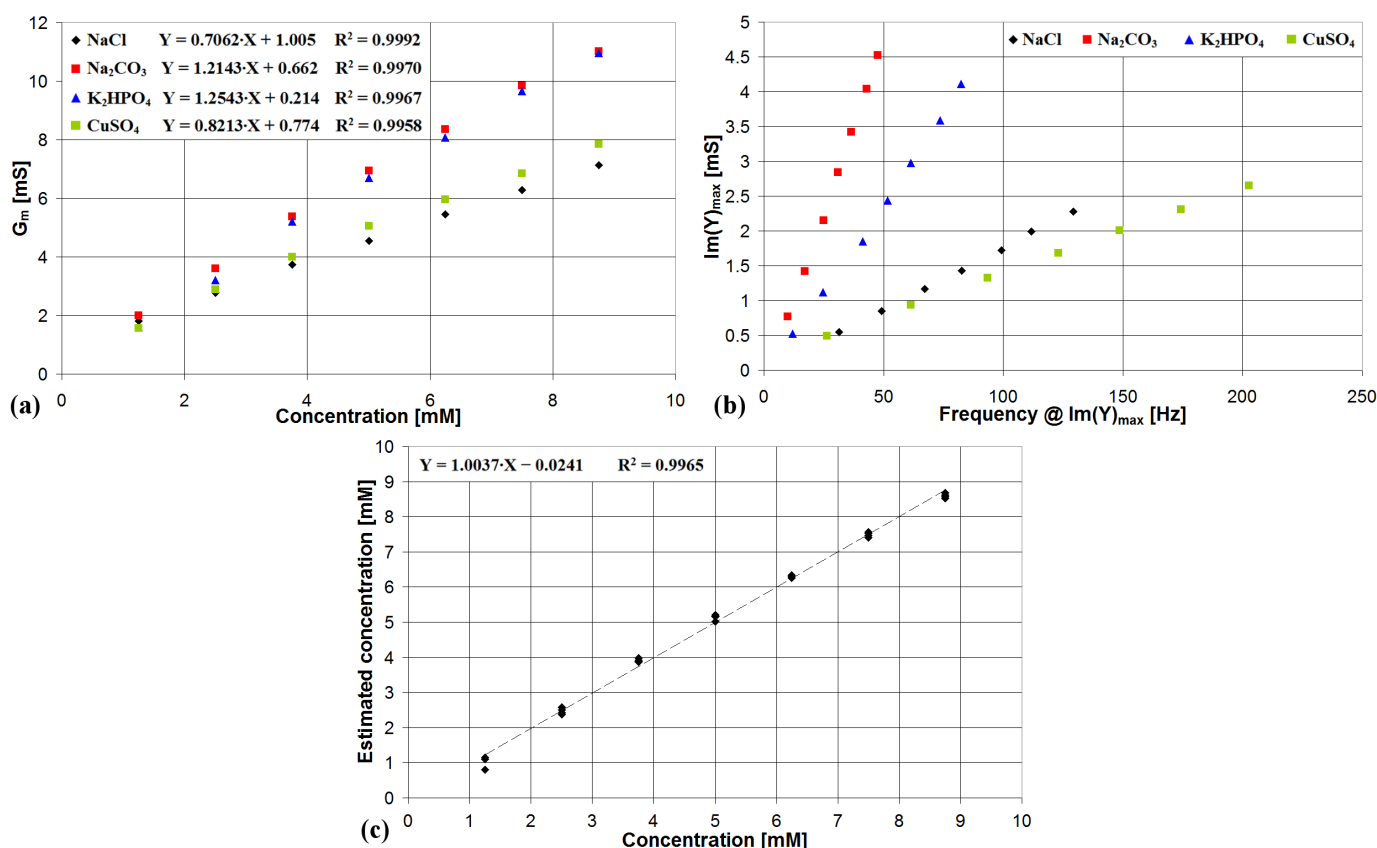
$$\text{Re}(Y) = \frac{\text{Re}(Z)}{|Z|^2} = \frac{\frac{1}{G_m} + \frac{\cos\left(\frac{\alpha\pi}{2}\right)}{Q\omega^\alpha}}{\left(\frac{1}{G_m}\right)^2 + \left(\frac{1}{Q\omega^\alpha}\right)^2 + \frac{2\cos\left(\frac{\alpha\pi}{2}\right)}{G_m Q\omega^\alpha}} \quad (7)$$

$$\text{Im}(Y) = -\frac{\text{Im}(Z)}{|Z|^2} = -\frac{\frac{\sin\left(\frac{\alpha\pi}{2}\right)}{Q\omega^\alpha}}{\left(\frac{1}{G_m}\right)^2 + \left(\frac{1}{Q\omega^\alpha}\right)^2 + \frac{2\cos\left(\frac{\alpha\pi}{2}\right)}{G_m Q\omega^\alpha}} \quad (8)$$

The measured spectra have been fitted with the proposed model and the results are presented in Table 2 for all the saline solutions and their concentrations. The presented data show that G_m is

Table 2. Electrical parameters of the tested saline solutions determined by fitting the measured impedance spectra with the proposed model.

Concentration (mM)	NaCl			Na ₂ CO ₃			K ₂ HPO ₄			CuSO ₄		
	G _m (mS)	Q (μF)	α	G _m (mS)	Q (μF)	α	G _m (mS)	Q (μF)	α	G _m (mS)	Q (μF)	α
1.25	1.81	47.4	0.689	2.01	62.8	0.836	1.60	64.5	0.742	1.57	40.7	0.715
2.5	2.77	49.6	0.701	3.61	68.3	0.849	3.20	63.9	0.776	2.89	36.3	0.735
3.75	3.74	51.5	0.709	5.37	70.1	0.859	5.21	65.4	0.788	3.99	34.5	0.745
5	4.55	52.6	0.713	6.94	69.1	0.875	6.69	64.6	0.802	5.05	34.8	0.748
6.25	5.46	53.8	0.717	8.35	72.1	0.874	8.07	64.9	0.810	5.96	33.9	0.755
7.5	6.28	55.8	0.721	9.85	73.4	0.875	9.66	65.4	0.814	6.86	34.8	0.755
8.75	7.14	55.6	0.724	11.0	74.8	0.876	10.9	65.6	0.819	7.85	35.4	0.756

**Fig. 6.** Measured electrical conductance vs molar concentration for different saline solutions (a); scatter plot of peak value of $Im(Y)$ plotted vs frequency of maximum value for different saline solutions (b); scatter plot of estimated molar concentration vs real molar concentration (c).

the parameter featuring the best correlation with the solution concentration.

Fig. 6 (a) shows G_m vs. solution concentration for all tested saline solutions. A very good linear correlation exists between the two variables ($R^2 > 0.99$). Two different groups can be identified, since for the same concentration Na₂CO₃ and K₂HPO₄ exhibit G_m values higher than NaCl and CuSO₄. This can be ascribed to the different ion concentration of the tested solutions: in fact, for the same molar concentration, Na₂CO₃ and K₂HPO₄ are characterized by double ion concentration with respect to NaCl and CuSO₄.

The electrical conductance, that can also be estimated with $Re(Y)$ at high frequency (10 KHz), is not sufficient to discriminate the different types of saline solutions but can be used to estimate the solution concentration if the particular saline solution is known. Repeatability tests carried out in triplicate on each sample have shown that the G_m dispersion (maximum deviation among the three measurements) is always $< 40 \mu S$.

The discrimination of the different saline solutions is however possible by analyzing the impedance spectrum on the full frequency range. As

can be seen in Fig. 5 (b) the imaginary component of Y ($\text{Im}(Y)_{\max}$) features a local maximum at a frequency f_{\max} . Then,

$$f_{\max} = \frac{1}{2\pi} e^{\frac{1}{\alpha} \ln\left(\frac{G_m}{Q}\right)} \quad (9)$$

$$\text{Im}(Y)_{\max} = \frac{G_m}{2} \cdot \frac{\sin\left(\frac{\alpha\pi}{2}\right)}{1 + \cos\left(\frac{\alpha\pi}{2}\right)} \quad (10)$$

Equations 9 and 10 have been used to calculate f_{\max} and $\text{Im}(Y)_{\max}$ for all tested samples using the equivalent electrical model parameters (G_m , Q , α) estimated from the measured EIS spectrum.

Fig. 6 (b) shows a scatter plot of $\text{Im}(Y)_{\max}$ vs f_{\max} for all tested saline solutions and concentrations. As can be seen, the couple of parameters f_{\max} and $\text{Im}(Y)_{\max}$ represent an electrical fingerprint allowing to discriminate the different solutions.

The scatter plot of Fig. 6 (b) has been divided in different regions corresponding to the different saline solutions. For each type of solution the linear regression line expressing $\text{Im}(Y)_{\max}$ as function of f_{\max} has been calculated. Then the scatter plot has been divided in four different regions so that the distance of the regression line of each type of solution from the corresponding region edge is maximized (in the sense in mean square). The results of the application of such clustering approach is the following: if $\text{Im}(Y)_{\max} > 0.065f_{\max}$ the solution is classified as Na_2CO_3 ; else if $\text{Im}(Y)_{\max} > 0.03f_{\max}$ is K_2HPO_4 ; else if $\text{Im}(Y)_{\max} > 0.0155f_{\max}$ it is NaCl ; else it is CuSO_4 .

Using this classification algorithm, all samples are correctly classified except for the lowest concentration (1.25 mM) of CuSO_4 that is misclassified as NaCl .

Once the solution type is determined, the solution concentration can be estimated from the measured value of G_m by using the calibration equations in the inset of Fig. 6 (a).

Estimated solution concentrations are plotted vs the real ones in Fig. 6 (c) and the high determination coefficient ($R^2 = 0.9965$) clearly indicates the good accuracy of the method of this work, independently of the type of saline solution.

4. Conclusions

A new low-cost portable electronic system for Electrical Impedance Spectroscopy (EIS) of liquid and semi-liquid media has been presented. The proposed system is composed of a polylactic acid structure realized with a 3D printer and an ad-hoc designed electronic board for electrical

measurements, as well as for data transfer to a laptop PC by USB port for data filing and further processing.

The system has been tested with four different saline solutions (NaCl , Na_2CO_3 , K_2HPO_4 and CuSO_4) and the results have shown that EIS spectrum provides an electrical fingerprint of the sample under test capable to discriminate the different types of solutions and to estimate their concentration with very good accuracy.

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